



# +3.3V, 2.5Gbps Low-Power Laser Driver

## General Description

The MAX3273 is a compact, low-power laser driver for applications up to 2.7Gbps. The device uses a single +3.3V supply and typically consumes 30mA. The bias and modulation current levels are programmed by external resistors. An automatic power-control (APC) loop is incorporated to maintain a constant average optical power over temperature and lifetime. The laser driver is fabricated using Maxim's in-house second generation SiGe process.

The MAX3273 accepts differential CML-compatible clock and data input signals. Inputs are self-biased to allow AC-coupling. An input data-retiming latch can be enabled to reject input jitter if a clock signal is available.

The driver can provide bias current up to 100mA and modulation current up to 60mA<sub>p-p</sub> with typical (20% to 80%) edge speeds of 59ps. A failure-monitor output is provided to indicate when the APC loop is unable to maintain average optical power. The MAX3273 is available in a 4mm x 4mm, 24-pin QFN package, as well as in die form.

## Applications

SONET OC-48 and SDH STM-16  
Transmission Systems  
Add/Drop Multiplexers  
Digital Cross-Connects  
2.5Gbps Optical Transmitters

## Features

- ◆ 30mA Power-Supply Current
- ◆ Single +3.3V Power Supply
- ◆ Up to 2.7Gbps (NRZ) Operation
- ◆ Automatic Average Power Control with Failure Monitor
- ◆ Programmable Modulation Current from 5mA to 60mA
- ◆ Programmable Bias Current from 1mA to 100mA
- ◆ Typical Fall Time of 59ps
- ◆ Selectable Data Retiming Latch
- ◆ Complies with ANSI, ITU, and Bellcore SDH/SONET Specifications

## Ordering Information

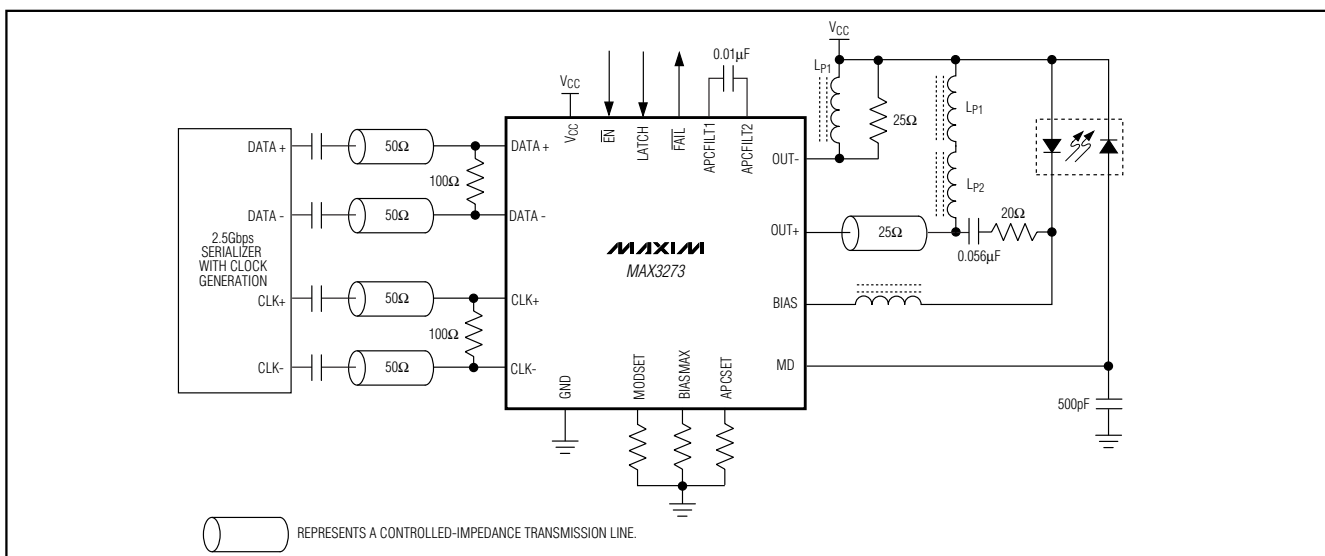
PART	TEMP. RANGE	PIN-PACKAGE
MAX3273EGG	-40°C to +85°C	24 QFN* (4mm x 4mm)
MAX3273E/D	-40°C to +85°C	Dice**

\* Exposed pad

\*\* Dice are designed to operate from  $T_A = -40^\circ\text{C}$  to  $+85^\circ\text{C}$ , but are tested and guaranteed at  $T_A = +25^\circ\text{C}$  only.

Pin Configuration appears at end of data sheet.

## Typical Applications Circuit



† Covered by U.S. patent number 5,883,910.

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## ABSOLUTE MAXIMUM RATINGS

Supply Voltage, $V_{CC}$ .....	-0.5V to +6.0V	Continuous Power Dissipation ( $T_A = +85^\circ\text{C}$ )	
Current into BIAS, OUT+, OUT- .....	-20mA to +150mA	24-Pin QFN (derate 274mW/ $^\circ\text{C}$ above +85 $^\circ\text{C}$ ) .....	1781mW
Current into MD .....	-5mA to +5mA	Storage Temperature Range .....	-55 $^\circ\text{C}$ to +150 $^\circ\text{C}$
Voltage at DATA+, DATA-, CLK+, CLK-, LATCH, $\overline{\text{EN}}$ , FAIL.....	-0.5V to ( $V_{CC} + 0.5\text{V}$ )	Operating Junction Temperature .....	-55 $^\circ\text{C}$ to +150 $^\circ\text{C}$
Voltage at MODSET, BIASMAX, APCSET, APCFLT1, APCFLT2.....	-0.5V to +3.0V	Die Attach Temperature (die) .....	+400 $^\circ\text{C}$
Voltage at BIAS .....	+1.0V to ( $V_{CC} + 1.5\text{V}$ )	Lead Temperature (soldering, 10s) .....	+300 $^\circ\text{C}$
Voltage at OUT+, OUT-.....	+1.5V to ( $V_{CC} + 1.5\text{V}$ )		
Current into $\overline{\text{FAIL}}$ .....	-10mA to +10mA		

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## DC ELECTRICAL CHARACTERISTICS

( $V_{CC} = +3.14\text{V}$  to +3.6V,  $T_A = -40^\circ\text{C}$  to +85 $^\circ\text{C}$ . Typical values are at  $V_{CC} = +3.3\text{V}$ ,  $I_{\text{BIAS}} = 60\text{mA}$ ,  $I_{\text{MOD}} = 30\text{mA}$ ,  $T_A = +25^\circ\text{C}$ , unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Supply Current	$I_{CC}$	Excluding $I_{\text{BIAS}}$ and $I_{\text{MOD}}$		30	45	mA
Bias-Current Range	$I_{\text{BIAS}}$	Voltage on BIAS pin ( $V_{\text{BIAS}} = V_{CC} - 1.6\text{V}$ )	1		100	mA
Bias Off-Current		$\overline{\text{EN}} = \text{high}$ (Note 2), $V_{\text{BIAS}} \leq 2.6\text{V}$			0.2	mA
Bias-Current Stability		APC open loop (Note 3)	$I_{\text{BIAS}} = 100\text{mA}$		61	ppm/ $^\circ\text{C}$
			$I_{\text{BIAS}} = 1\text{mA}$		198	
Bias-Current Absolute Accuracy		APC open loop (Note 4)	-15		+15	%
Differential Input Voltage	$V_{\text{ID}}$	Figure 1	0.2		1.6	Vp-p
Common-Mode Input Voltage	$V_{\text{ICM}}$		$V_{CC} - 1.49$	$V_{CC} - 1.32$	$V_{CC} - V_{\text{ID}}/4$	V
TTL Input High Voltage	$V_{\text{IH}}$		2.0			V
TTL Input Low Voltage	$V_{\text{IL}}$				0.8	V
TTL Output High	$V_{\text{OH}}$	Sourcing 50 $\mu\text{A}$	2.4			V
TTL Output Low	$V_{\text{OL}}$	Sinking 100 $\mu\text{A}$			0.4	V
MD Voltage					1.6	V
Monitor Diode DC-Current Range	$I_{\text{MD}}$	(Note 3)	18		1000	$\mu\text{A}$
Monitor-Diode Bias Set point Stability		$I_{\text{MD}} = 1000\mu\text{A}$	-480	83	+480	ppm/ $^\circ\text{C}$
		$I_{\text{MD}} = 18\mu\text{A}$	-480	159	+480	
Monitor-Diode Bias Absolute Accuracy			-15		+15	%

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## AC ELECTRICAL CHARACTERISTICS

( $V_{CC} = +3.14V$  to  $+3.6V$ ,  $T_A = -40^{\circ}C$  to  $+85^{\circ}C$ . Typical values are at  $V_{CC} = +3.3V$ ,  $I_{BIAS} = 60mA$ ,  $I_{MOD} = 30mA$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.) (Notes 5, 6)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Modulation-Current Range	$I_{MOD}$	(Note 3)	5		60	mA
Modulation Off-Current		$\overline{EN} = \text{High}$			0.2	mA
Modulation-Current Stability		$I_{MOD} = 60mA$	-480	64	+480	ppm/ $^{\circ}C$
		$I_{MOD} = 5mA$	-480	34	+480	
Modulation-Current Absolute Accuracy		(Note 4)	-15		+15	%
Output Current Rise Time	$t_R$	20% to 80% (Note 7)		52	87	ps
Output Current Fall Time	$t_F$	20% to 80% (Note 7)		59	104	ps
Output Overshoot/Undershoot	$\delta$	(Note 7)		15		%
Enable and Startup Delay		APC open loop		364		ns
Maximum Consecutive Identical Digits			80			bits
Pulse-Width Distortion	PWD	(Notes 7, 8)		3	45	ps
Random Jitter				1.0	1.5	psRMS
Input Latch Setup Time	$T_{SU}$	LATCH = high (Figure 1)		75	150	ps
Input Latch Hold Time	$T_{HD}$	LATCH = high (Figure 1)		0	50	ps

**Note 1:** Specifications at  $-40^{\circ}C$  are guaranteed by design and characterization. Dice are tested at  $T_A = +25^{\circ}C$  only.

**Note 2:** Both the bias and modulation currents will be switched off if any of the current set pins are grounded.

**Note 3:** Guaranteed by design and characterization.

**Note 4:** Accuracy refers to part-to-part variation.

**Note 5:** AC characterization was performed by using the circuit in Figure 2.

**Note 6:** AC characteristics are guaranteed by design and characterization, and measured using a 2.5Gbps  $2^{13} - 1$  PRBS input data pattern with 80 consecutive zeros and 80 consecutive ones added.

**Note 7:** Measured using a 2.5Gbps repeating 0000 1111 pattern.

**Note 8:**  $PWD = (\text{wide pulse} - \text{narrow pulse})/2$ .

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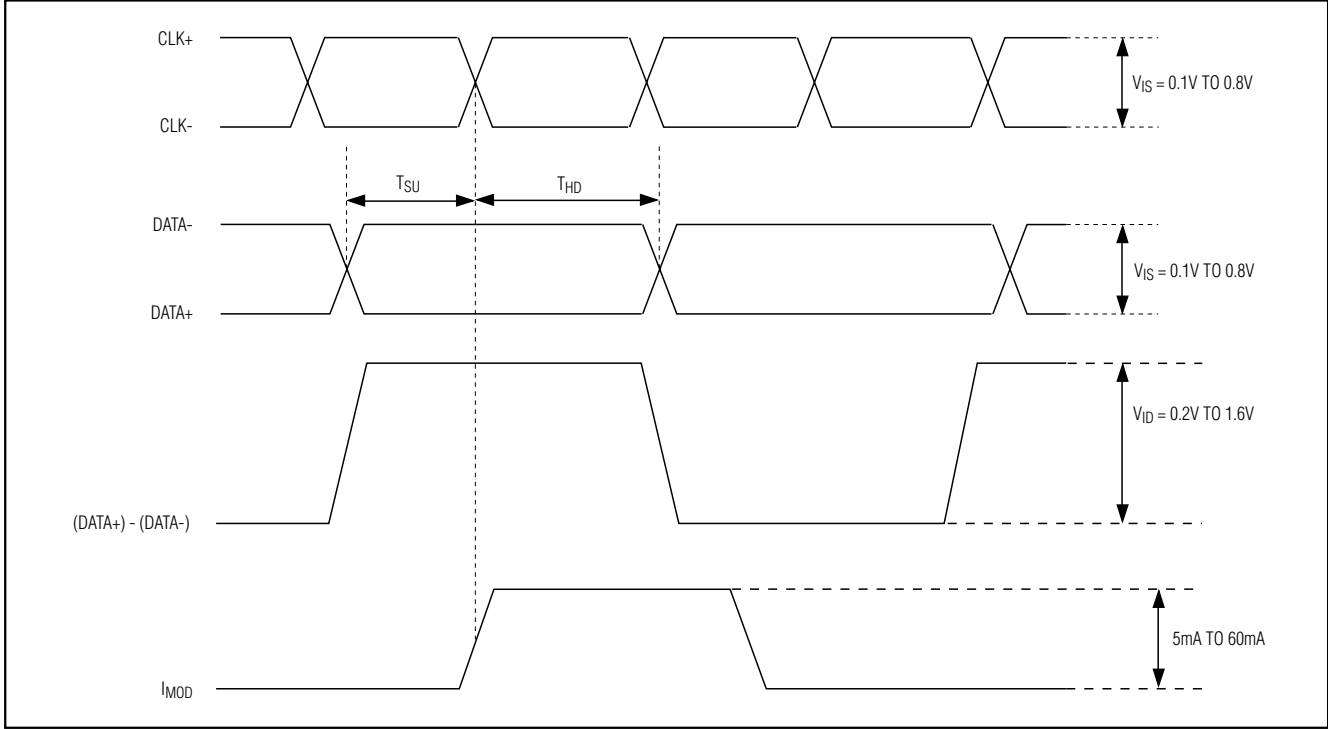


Figure 1. Required Input Signal and Setup/Hold-Time Definition

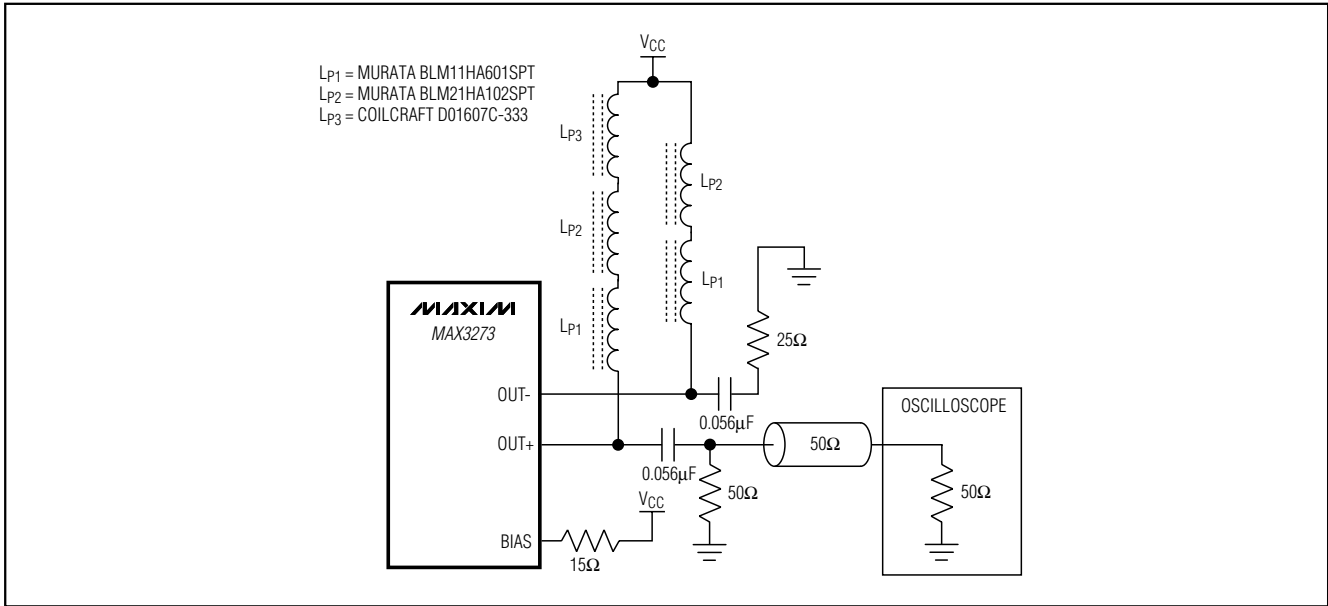


Figure 2. Output Termination for Characterization

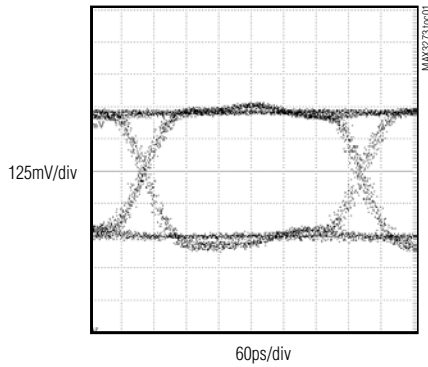
# +3.3V, 2.5Gbps Low-Power Laser Driver

**MAX3273**

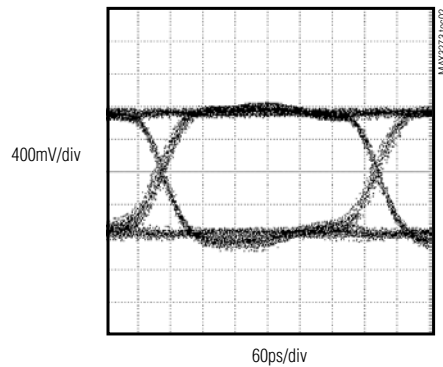
## Typical Operating Characteristics

( $V_{CC} = 3.3V$ ,  $T_A = +25^\circ C$ , unless otherwise noted.)

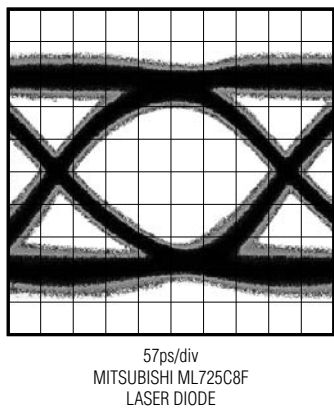
**ELECTRICAL EYE DIAGRAM**  
( $I_{MOD} = 20mA$ ,  $2^{13} - 1$  80CID)



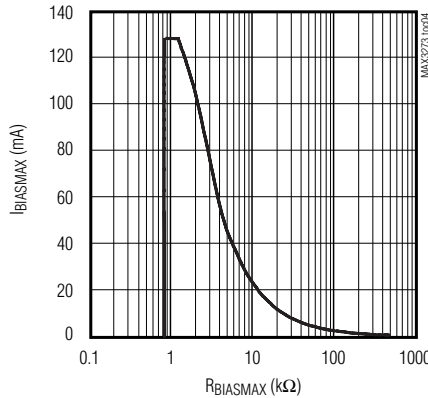
**ELECTRICAL EYE DIAGRAM**  
( $I_{MOD} = 60mA$ ,  $2^{13} - 1$  80CID)



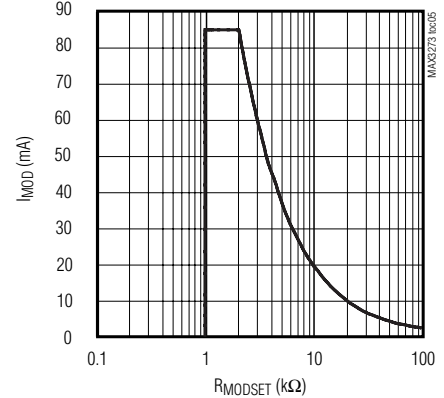
**OPTICAL EYE DIAGRAM**  
(2.488Gbps, 1300nm FP LASER,  
1.87GHz FILTER)



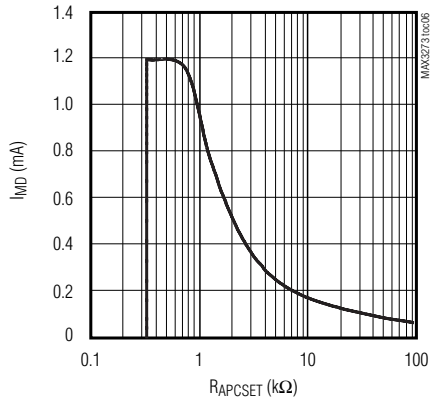
**I<sub>BIASMAX</sub> vs. R<sub>BIASMAX</sub>**



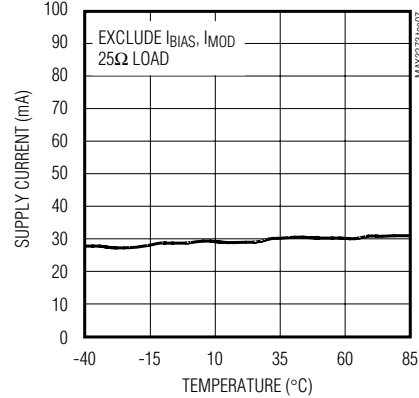
**I<sub>MOD</sub> vs. R<sub>MODSET</sub>**



**I<sub>MD</sub> vs. R<sub>APCSET</sub>**



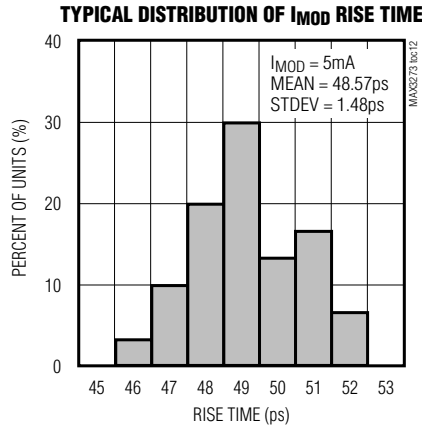
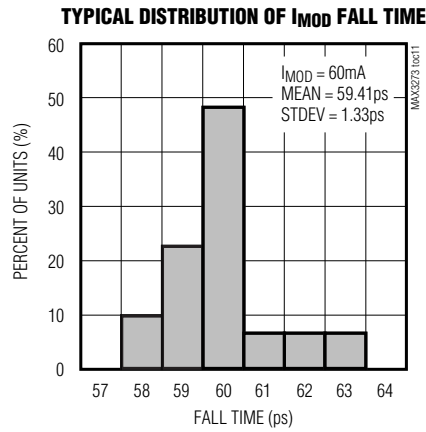
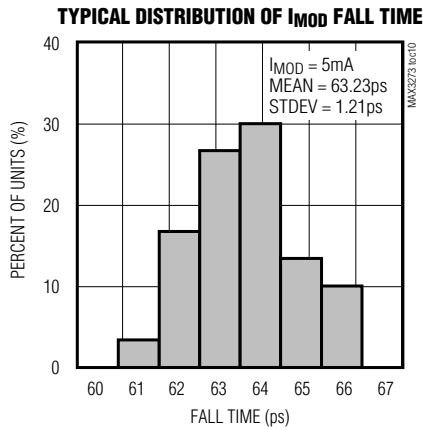
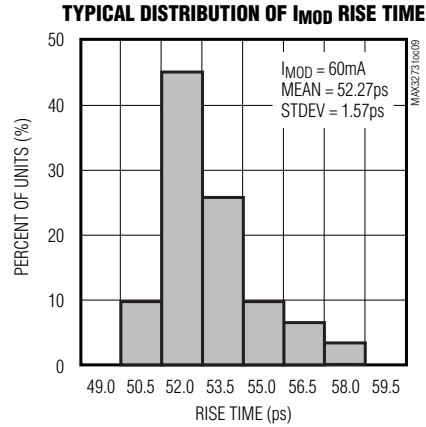
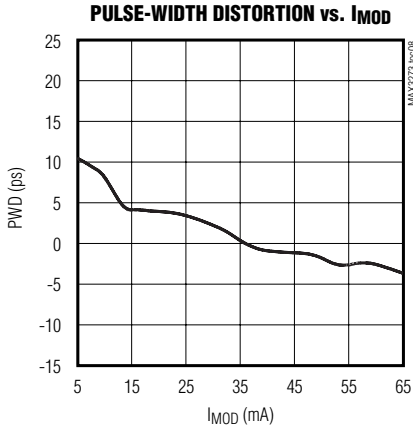
**SUPPLY CURRENT vs. TEMPERATURE**



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## Typical Operating Characteristics (continued)

( $V_{CC} = 3.3V$ ,  $T_A = +25^\circ C$ , unless otherwise noted.)



# +3.3V, 2.5Gbps Low-Power Laser Driver

## Pin Description

**MAX3273**

PIN	NAME	FUNCTION
1, 4, 13, 15, 18	V <sub>CC</sub>	Power-Supply Voltage
2	DATA+	Noninverting Data Input, with On-Chip Biasing
3	DATA-	Inverting Data Input, with On-Chip Biasing
5	CLK+	Noninverting Clock Input for Data Retiming, with On-Chip Biasing
6	CLK-	Inverting Clock Input for Data Retiming, with On-Chip Biasing
7, 9, 12	GND	Ground
8	LATCH	Data Retiming Enable Input, Active-High. Retiming disabled when floating or pulled low.
10	$\overline{\text{EN}}$	TTL/CMOS Enable Input. Low for normal operation. Float or pull high to disable laser bias and modulation currents. Internal 100k $\Omega$ pullup to V <sub>CC</sub> .
11	MODSET	A resistor connected from this pin to ground sets the desired modulation current.
14	BIAS	Laser Bias Current Output. Connect to the laser via an inductor.
16	OUT+	Positive Modulation-Current Output. I <sub>MOD</sub> flows into this pin when input data is high.
17	OUT-	Negative Modulation-Current Output. Current flows into this pin when input data is low. Connect to load equivalent to that on OUT+ to maintain differential output balance.
19	MD	Monitor Diode Input. Connect this pin to the anode of the monitor diode. Leave floating for open-loop operation.
20	APCFILT1	A capacitor between APCFILT1 and APCFILT2 sets the dominant pole of the APC feedback loop (C <sub>APCFILT</sub> = 0.01 $\mu$ F). Ground APCFILT1 for open-loop operation.
21	APCFILT2	See above
22	$\overline{\text{FAIL}}$	TTL/CMOS Failure Output, Active-Low. Indicates APC failure when low.
23	APCSET	A resistor connected from this pin to ground sets the desired average optical power. Connect a 100k $\Omega$ resistor to GND for open-loop operation.
24	BIASMAX	A resistor connected from this pin to ground sets the maximum bias current. The APC function can subtract current from this maximum value, but cannot add to it. For open-loop operation, this pin sets the laser bias current.
EP	EXPOSED PAD	Ground. Solder this pad to ground.
CP	CORNER PINS	Ground. These pins must be soldered to the circuit board ground for proper performance.

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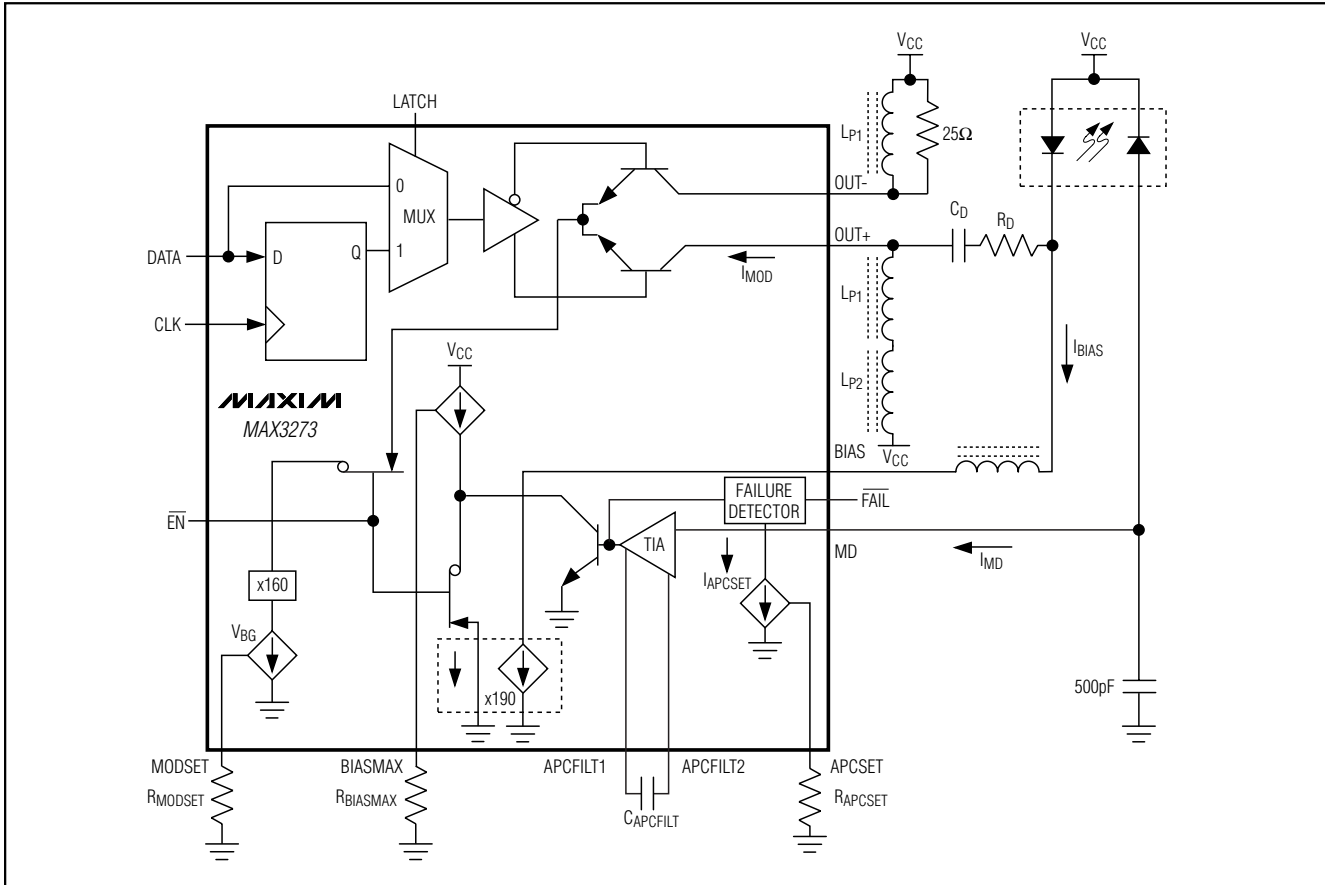


Figure 3. Functional Diagram

## Detailed Description

The MAX3273 laser driver consists of two main parts: a high-speed modulation driver and a laser-biasing block with automatic power control (APC). The circuit design is optimized for both high-speed and low-voltage (+3.3V) operation. To minimize the jitter of the input signal at speeds as high as 2.7Gbps, the device accepts a differential CML clock signal for data retiming. When LATCH is high, the input data is synchronized by the clock signal. When LATCH is low, the input data is directly applied to the output stage.

The output stage is composed of a high-speed differential pair and a programmable modulation current source. Since the modulation output drives a maximum current of 60mA into the laser with an edge speed of 59ps, large transient voltage spikes can be generated (due to the parasitic inductance of the laser). These transients and the laser-forward voltage leave insuffi-

cient headroom for the proper operation of the laser driver if the modulation output is DC-coupled to the laser diode. To solve this problem, the MAX3273's modulation output is AC-coupled to the cathode of a laser diode. An external pullup inductor is necessary to DC-bias the modulation output at  $V_{CC}$ . Such a configuration isolates laser-forward voltage from the output circuitry and the supply voltage  $V_{CC}$ . A simplified functional diagram is shown in Figure 3.

The MAX3273 modulation output is optimized for driving a 25Ω load. Modulation current swings of 75mA are possible, but due to minimum power-supply and jitter requirements at 2.5Gbps, the specified maximum modulation current is limited to 60mA. To interface with the laser diode, a damping resistor ( $R_D$ ) is required for impedance matching. An RC shunt network may also be necessary to compensate for the laser-diode parasitic inductance, thereby improving the optical output



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ringing and duty-cycle distortion. Please refer to the Maxim application note HFAN 02.0, *Interfacing Maxim Laser Drivers with Laser Diodes*, for more information.

At the data rate of 2.5Gbps, any capacitive load at the cathode of a laser diode will degrade the optical output performance. Since the BIAS output is directly connected to the laser cathode, the parasitic capacitance associated with this pin is minimized by using an inductor to isolate the BIAS pin from the laser cathode.

## Automatic Power Control (APC)

To maintain constant average optical power, the MAX3273 incorporates an APC loop to compensate for the changes in laser threshold current over temperature and lifetime. A back-facet photodiode mounted in the laser package is used to convert the optical power into a photocurrent. The APC loop adjusts the laser bias current so that the monitor current is matched to a reference current set by R<sub>APCSET</sub>. The time constant of the APC loop is determined by an external capacitor (CAPCFILT). To minimize the pattern-dependent jitter associated with the APC loop-time constant, and to guarantee loop stability, the recommended value for CAPCFILT is 0.01μF.

When the APC loop is functioning, the maximum allowable bias current is set by an external resistor, R<sub>BIASMAX</sub>. An APC failure flag ( $\overline{\text{FAIL}}$ ) is asserted low when the bias current can no longer be adjusted to achieve the desired average optical power.

APC closed-loop operation requires the user to set three currents with external resistors connected between ground and BIASMAX, MODSET, and APCSET (see Figure 3). Detailed guidelines for these resistor settings are described in the *Design Procedure* section.

## Open-Loop Operation

If necessary, the MAX3273 is fully operational without APC. To disable the APC loop, ground the APCFLT1 pin. In this case, the laser current is directly set by two external resistors connected from ground to BIASMAX and MODSET. See the *Design Procedure* section for more details on open-loop operation.

## Optional Data Input Latch

To minimize jitter in the input data, a synchronous differential clock signal should be connected to the CLK+ and CLK- inputs. When the LATCH control input is tied high, the input data is retimed on the rising edge of CLK+. If LATCH is tied low or left floating, the retiming function is disabled and the input data is directly connected to the output stage. When this latch function is not used, connect CLK+ to V<sub>CC</sub> and leave CLK- unconnected.

## Output Enable

The MAX3273 incorporates a TTL/CMOS input to enable the output. When  $\overline{\text{EN}}$  is low, the modulation and bias outputs are enabled. When  $\overline{\text{EN}}$  is high or floating, both the bias and modulation currents are off. The typical enable time is 364ns and the typical disable time is 27ns when the bias is operated open-loop.

## Slow-Start

For laser safety reasons, the MAX3273 incorporates a slow-start circuit that provides a delay of 364ns for enabling a laser diode.

## APC Failure Monitor

The MAX3273 provides an APC failure monitor (TTL/CMOS) to indicate an APC loop tracking failure.  $\overline{\text{FAIL}}$  is asserted low when the APC loop can no longer regulate the bias current to maintain the desired monitor diode current.  $\overline{\text{FAIL}}$  will assert low when the APC loop is disabled.

## Short-Circuit Protection

The MAX3273 provides short-circuit protection for the modulation and bias current sources. If BIASMAX, MODSET, or APCSET are shorted to ground, the bias and modulation output will turn off.

## Design Procedure

When designing a laser transmitter, the optical output is usually expressed in terms of average power and extinction ratio. Table 1 gives relationships that are helpful in converting between the optical average power and the modulation current. These relationships are valid if the mark density and duty cycle of the optical waveform are 50%.

## Programming the Modulation Current

For a given laser power P<sub>AVG</sub>, slope efficiency η, and extinction ratio (r<sub>e</sub>), the modulation current can be calculated using Table 1. See the I<sub>MOD</sub> vs. R<sub>MODSET</sub> graph in the *Typical Operating Characteristics* and select the value of R<sub>MODSET</sub> that corresponds to the required current at +25°C. The equation below provides a derivation of the modulation current using Table 1.

$$I_{\text{MOD}} = 2 \times \frac{P_{\text{AVE}}}{\eta} \times \frac{r_e - 1}{r_e + 1}$$

## Programming the Bias Current with APC Disabled

When using the MAX3273 in open-loop operation, the bias current is determined by the R<sub>BIASMAX</sub> resistor. To select this resistor, see the I<sub>BIASMAX</sub> vs. R<sub>BIASMAX</sub> graph in the *Typical Operating Characteristics* and select the

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value of  $R_{BIASMAX}$  that corresponds to the required  $I_{BIASMAX}$  at +25°C. Ground the APCFILT1 pin for open-loop operation.

### Programming the Bias Current with APC Enabled

When the MAX3273's APC feature is used, program the average optical power by adjusting the APCSET resistor. To select this resistor, determine the desired monitor current to be maintained over temperature and life. See the  $I_{MD}$  vs.  $R_{APCSET}$  graph in the *Typical Operating Characteristics* and select the value of  $R_{APCSET}$  that corresponds to the required current.

When using the MAX3273 in closed-loop operation, the  $R_{BIASMAX}$  resistor sets the maximum bias current available to the laser diode over temperature and life. The APC loop can subtract from this maximum value but cannot add to it. See the  $I_{BIASMAX}$  vs.  $R_{BIASMAX}$  graph in the *Typical Operating Characteristics* and select the value of  $R_{BIASMAX}$  that corresponds to the end-of-life bias current at +85°C.

### Interfacing with Laser Diodes

To minimize optical output aberrations caused by signal reflections at the electrical interface to the laser diode, a series-damping resistor ( $R_D$ ) is required (see *Typical Application Circuit*). Additionally, the MAX3273 outputs are optimized for a 25Ω load. Therefore, the series combination of  $R_D$  and  $R_L$  (where  $R_L$  represents the laser-diode resistance) should equal 25Ω. Typical values for  $R_D$  are 18Ω to 23Ω. For best performance, a bypass capacitor (0.01μF typical) should be placed as close as possible to the anode of the laser diode. Depending on the exact characteristics of the laser diode and PC board layout, a resistor ( $R_P$ ) of 50Ω to 100Ω in parallel with pullup inductor  $L_{P1}$  can be useful in damping overshoot and ringing in the optical output.

In some applications (depending on laser-diode parasitic inductance), an RC-shunt network between the laser cathode and ground will help minimize optical output aberrations. Starting values for most coaxial lasers are  $R = 75\Omega$  in series with  $C = 3.3pF$ . These values should be experimentally adjusted until the optical output waveform is optimized.

### Pattern-Dependent Jitter

When transmitting NRZ data with long strings of consecutive identical digits (CIDs), LF droop can occur and contribute to pattern-dependent jitter (PDJ). To minimize this PDJ, three external components must be properly chosen: capacitor  $C_{APCFILT}$ , which dominates the APC loop time constant; pullup inductor  $L_P$ ; and AC-coupling capacitor  $C_D$ .

To filter out noise effects and guarantee loop stability, the recommended value for  $C_{APCFILT}$  is 0.01μF. This results in an APC loop bandwidth of 100kHz or a time constant of 15μs. As a result, the PDJ associated with an APC loop time constant can be ignored.

The time constant associated with the output pullup inductor ( $L_P \approx L_{P2}$ ) and the AC-coupling capacitor ( $C_D$ ) will impact the PDJ. For such a second-order network, the PDJ will be dominated by  $L_P$  due to the low frequency cutoff. For a data rate of 2.5Gbps, the recommended value for  $C_D$  is 0.056μF. During the maximum CID period, limit the peak voltage droop to less than 12% of the average (6% of the amplitude). The time constant can be estimated by:

$$12\% = 1 - e^{-\frac{t}{\tau_{LP}}}$$

$$\tau_{LP} = 7.8t$$

If  $\tau_{LP} = L_P/25\Omega$ , and  $t = 100UI \approx 40ns$ , then  $L_P = 7.8\mu H$ . To reduce the physical size of this element ( $L_P$ ), use of SMD ferrite beads is recommended (Figure 2). To achieve even greater immunity to droop, an optional third inductor, 33μH, can be used ( $L_{P3}$  in Figure 2).

### Input Termination Requirement

The MAX3273 data and clock inputs are CML compatible. However, it is not necessary to drive the IC with a standard CML signal. As long as the specified differential voltage swings are met, the MAX3273 will operate properly.

### Calculating Power Consumption

The junction temperature of the MAX3273 dice must be kept below +150°C at all times. The total power dissipation of the MAX3273 can be estimated by the following:

$$P = V_{CC} \times I_{CC} + (V_{CC} - V_f) \times I_{BIAS} + I_{MOD} \times (V_{CC} - 25 \times I_{MOD} / 2)$$

where  $I_{BIAS}$  is the maximum bias current set by  $R_{BIASMAX}$ ,  $I_{MOD}$  is the modulation current, and  $V_f$  is the typical laser forward voltage.

$$\text{Junction Temperature} = P(W) \times 37 \text{ (}^\circ\text{C/W)}$$

## Applications Information

An example of how to set up the MAX3273 follows.

### Select Laser

A communication-grade laser should be selected for 2.5Gbps/2.7Gbps applications. Assume the laser output average power is  $P_{AVG} = 0$ , minimum extinction ratio is  $r_e = 6.6$  (8.2dB), the operating temperature is -

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**Table 1. Optical Power Relations**

PARAMETER	SYMBOL	RELATION
Average Power	$P_{AVG}$	$P_{AVG} = (P_0 + P_1)/2$
Extinction Ratio	$r_e$	$r_e = P_1/P_0$
Optical Power of a "1"	$P_1$	$P_1 = 2P_{AVG}r_e/(r_e + 1)$
Optical Power of a "0"	$P_0$	$P_0 = 2P_{AVG}/(r_e + 1)$
Optical Amplitude	$P_{p-p}$	$P_{p-p} = P_1 - P_0 = 2P_{AVG}(r_e - 1)/(r_e + 1)$
Laser Slope Efficiency	$\eta$	$\eta = P_{p-p}/I_{MOD}$
Modulation Current	$I_{MOD}$	$I_{MOD} = P_{p-p}/\eta$
Threshold Current	$I_{TH}$	$P_0$ at $1 \geq I_{TH}$
Bias Current	$I_{BIAS}$	$I_{BIAS} \geq I_{TH} + I_{MOD}/2$
Laser-to-Monitor Transfer	$\rho_{MON}$	$I_{MD}/P_{AVG}$

**Note:** Assuming a 50% average input duty cycle and mark density.

40°C to +85°C, and the laser diode has the following characteristics:

Wavelength:  $\lambda = 1310\text{nm}$

Threshold Current:  $I_{TH} = 22\text{mA}$  at +25°C

Threshold Temperature Coefficient:  $\beta_{TH} = 1.3\%/^{\circ}\text{C}$

Laser to Monitor Transfer:  $\rho_{MON} = 0.2\text{A/W}$

Laser Slope Efficiency:  $\eta = 0.05\text{mW/mA}$  at +25°C

### Determine $R_{APCSET}$

The desired monitor diode current is estimated by  $I_{MD} = P_{AVG} \times \rho_{MON} = 200\mu\text{A}$ . The  $I_{MD}$  vs.  $R_{APCSET}$  graph in the *Typical Operating Characteristics* shows that  $R_{APCSET}$  should be 7.5k $\Omega$ .

### Determine $R_{MODSET}$

To achieve a minimum extinction ratio ( $r_e$ ) of 6.6 over temperature and lifetime, calculate the required extinction ratio at +25°C. Assuming  $r_e = 20$ , the peak-to-peak optical power  $P_{p-p} = 1.81\text{mW}$ , according to Table 1. The required modulation current is  $1.81\text{mW} / (0.05\text{mW/mA}) = 36.2\text{mA}$ . The  $I_{MOD}$  vs.  $R_{MODSET}$  graph

in the *Typical Operating Characteristics* shows that  $R_{MODSET}$  should be 5k $\Omega$ .

### Determine $R_{BIASMAX}$

Calculate the maximum threshold current ( $I_{TH(MAX)}$ ) at  $T_A = +85^{\circ}\text{C}$  and end of life. Assuming  $I_{TH(MAX)} = 50\text{mA}$ , the maximum bias current should be:  $I_{BIASMAX} = I_{TH(MAX)} + (I_{MOD} / 2)$ . In this example,  $I_{BIASMAX} = 68.1\text{mA}$ . The  $I_{BIASMAX}$  vs.  $R_{BIASMAX}$  graph in the *Typical Operating Characteristics* shows that  $R_{BIASMAX}$  should be 3.5k $\Omega$ .

### Interface Models

Figures 4 and 5 show simplified input and output circuits for the MAX3273 laser driver. If dice are used, replace package parasitic elements with bondwire parasitic elements.

### Wire Bonding Die

For high-current density and reliable operation, the MAX3273 uses gold metalization. Make connections to the die with gold wire only, using ball-bonding techniques. Wedge bonding is not recommended. Die-pad

## **+3.3V, 2.5Gbps Low-Power Laser Driver**

size is 4mils (100 $\mu$ m) square, and die thickness is 14mils (350 $\mu$ m).

### **Layout Considerations**

To minimize inductance, keep the connections between the MAX3273 output pins and laser diode as close as possible. Optimize the laser-diode performance by placing a bypass capacitor as close as possible to the laser anode. Use good high-frequency layout techniques and multilayer boards with uninterrupted ground planes to minimize EMI and crosstalk.

### **Laser Safety and IEC 825**

Using the MAX3273 laser driver alone does not ensure that a transmitter design is compliant with IEC 825. The entire transmitter circuit and component selections must be considered. Each customer must determine the level of fault tolerance required by their application, recognizing that Maxim products are not designed or authorized for use as components in systems intended for surgical implant into the body, for applications intended to support or sustain life, or for any other application where the failure of a Maxim product could create a situation where personal injury or death may occur.

### **Chip Information**

TRANSISTOR COUNT: 1672

PROCESS: SiGe

ISOLATED SUBSTRATE

# +3.3V, 2.5Gbps Low-Power Laser Driver

**MAX3273**

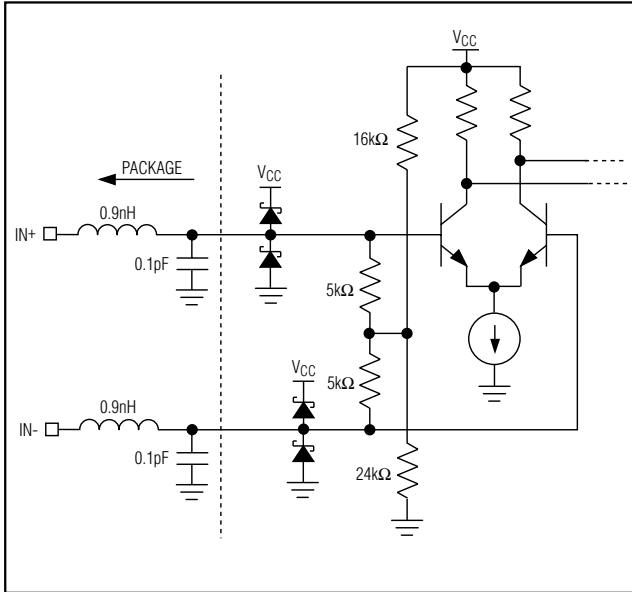


Figure 4. Simplified Input Circuit

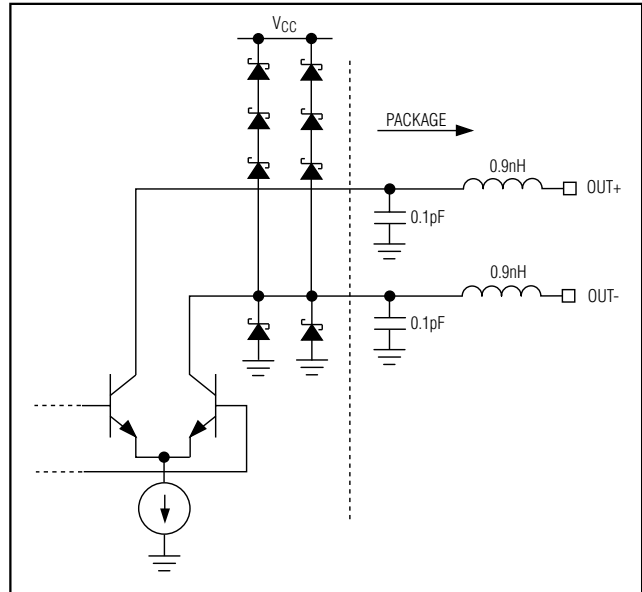
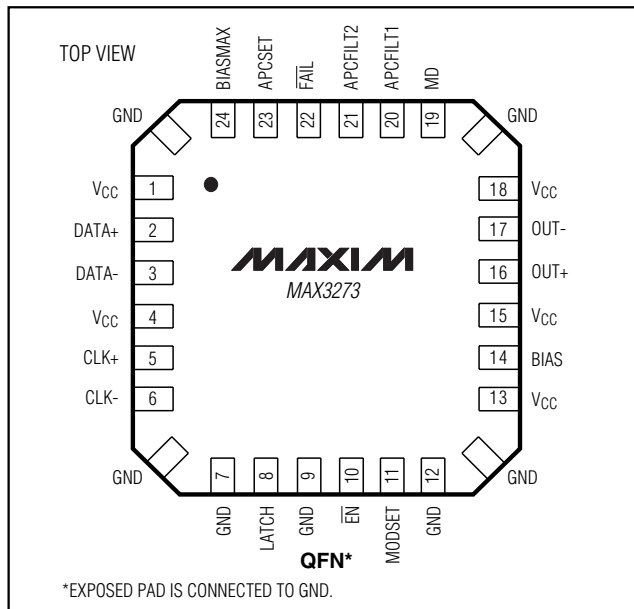
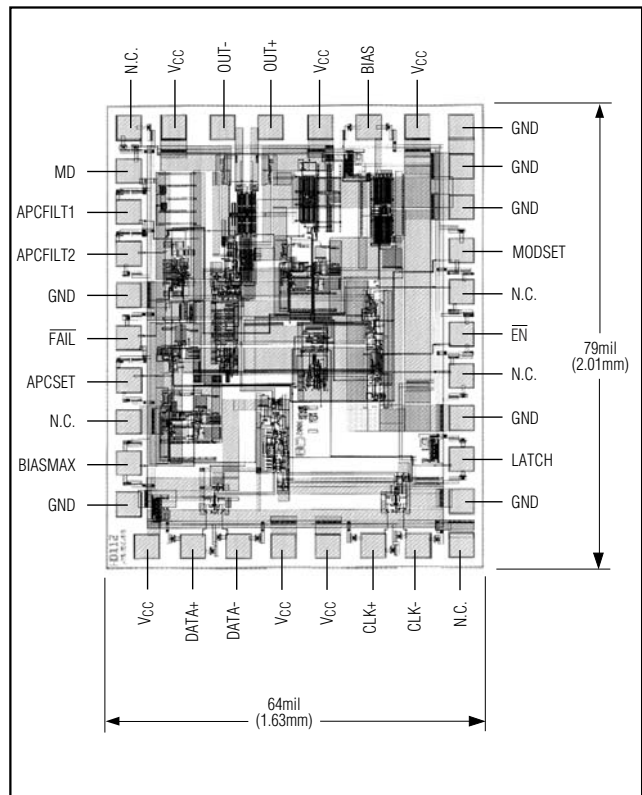


Figure 5. Simplified Output Circuit

## Pin Configuration

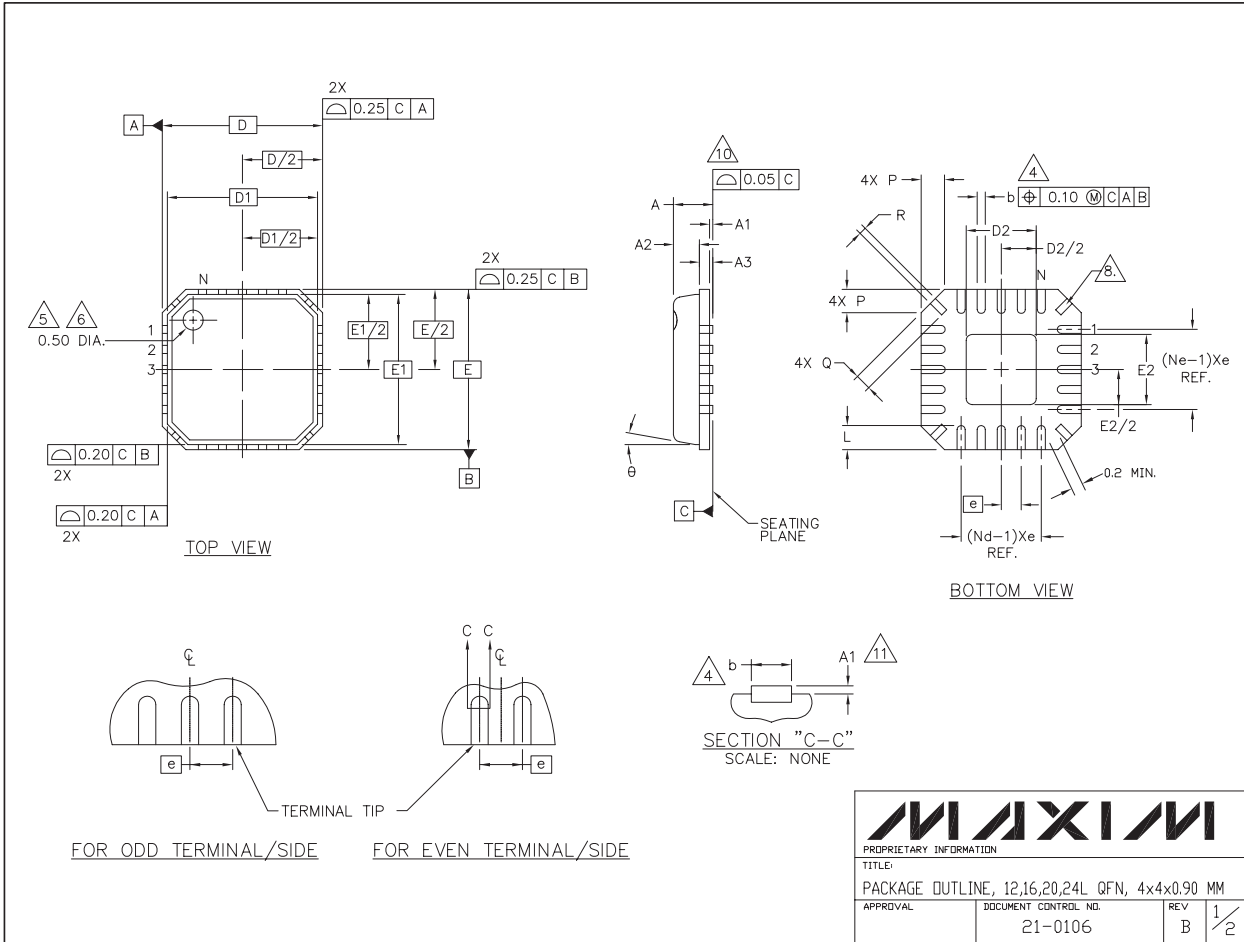


## Chip Topography



# +3.3V, 2.5Gbps Low-Power Laser Driver

## Package Information



<b>MAXIM</b>		
PROPRIETARY INFORMATION		
TITLE:		
PACKAGE OUTLINE, 12,16,20,24L QFN, 4x4x0.90 MM		
APPROVAL	DOCUMENT CONTROL NO.	REV
	21-0106	B 1/2

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## Package Information (continued)

MAX3273

### NOTES:

1. DIE THICKNESS ALLOWABLE IS 0.305mm MAXIMUM (.012 INCHES MAXIMUM)
2. DIMENSIONING & TOLERANCES CONFORM MUST TO ASME Y14.5M. - 1994.
3. N IS THE NUMBER OF TERMINALS.  
Nd IS THE NUMBER OF TERMINALS IN X-DIRECTION &  
Ne IS THE NUMBER OF TERMINALS IN Y-DIRECTION.
4. DIMENSION b APPLIES TO PLATED TERMINAL AND IS MEASURED BETWEEN 0.20 AND 0.25mm FROM TERMINAL TIP.
5. THE PIN #1 IDENTIFIER MUST BE EXISTED ON THE TOP SURFACE OF THE PACKAGE BY USING INDENTATION MARK OR OTHER FEATURE OF PACKAGE BODY.
6. EXACT SHAPE AND SIZE OF THIS FEATURE IS OPTIONAL.
7. ALL DIMENSIONS ARE IN MILLIMETERS.
8. THE SHAPE SHOWN ON FOUR CORNERS ARE NOT ACTUAL 1/0.
9. PACKAGE WARPAGE MAX 0.05mm.
10. APPLIED FOR EXPOSED PAD AND TERMINALS.  
EXCLUDE EMBEDDING PART OF EXPOSED PAD FROM MEASURING.
11. APPLIED ONLY FOR TERMINALS.
12. MEETS JEDEC MO220.

SYMBOL	COMMON DIMENSIONS			No. of Terminals
	MIN.	NOM.	MAX.	
A	—	0.85	1.00	
A1	0.00	0.01	0.05	11
A2	—	0.65	0.80	
A3	0.20 REF.			
D	4.00 BSC			
D1	3.75 BSC			
E	4.00 BSC			
E1	3.75 BSC			
θ	12°			
P	0.24	0.42	0.60	
R	0.13	0.17	0.23	

SYMBOL	PITCH VARIATION A			No. of Terminals	SYMBOL	PITCH VARIATION B			No. of Terminals	SYMBOL	PITCH VARIATION C			No. of Terminals	SYMBOL	PITCH VARIATION D			No. of Terminals
	MIN.	NOM.	MAX.			MIN.	NOM.	MAX.			MIN.	NOM.	MAX.			MIN.	NOM.	MAX.	
ⓐ	0.80 BSC				ⓐ	0.65 BSC				ⓐ	0.50 BSC				ⓐ	0.50 BSC			
N	12			3	N	16			3	N	20			3	N	24			3
Nd	3			3	Nd	4			3	Nd	5			3	Nd	6			3
Ne	3			3	Ne	4			3	Ne	5			3	Ne	6			3
L	0.50	0.60	0.75		L	0.50	0.60	0.75		L	0.50	0.60	0.75		L	0.30	0.40	0.55	
b	0.28	0.33	0.40	4	b	0.23	0.28	0.35	4	b	0.18	0.23	0.30	4	b	0.18	0.23	0.30	4
Q	0.30	0.40	0.65		Q	0.30	0.40	0.65		Q	0.30	0.40	0.65		Q	0.00	0.20	0.45	
D2	SEE EXPOSED PAD VARIATION: A, B				D2	SEE EXPOSED PAD VARIATION: A, B				D2	SEE EXPOSED PAD VARIATION: A, B				D2	SEE EXPOSED PAD VARIATION: A			
E2	SEE EXPOSED PAD VARIATION: A, B				E2	SEE EXPOSED PAD VARIATION: A, B				E2	SEE EXPOSED PAD VARIATION: A, B				E2	SEE EXPOSED PAD VARIATION: A			

SYMBOLS	D2			E2			NOTE	
	MIN	NOM	MAX	MIN	NOM	MAX		
EXPOSED PAD VARIATIONS	A	1.95	2.10	2.25	1.95	2.10	2.25	
	B	1.55	1.70	1.85	1.55	1.70	1.85	

EXAMPLE: WE CAN CALL VARIATION "BB" FOR 16 TERMINAL QFN WITH 1.70x1.70 mm NOMINAL EXPOSED PAD DIMENSION. THE FORMER ONE IN VARIATION IS FOR PITCH VARIATION AND THE LATTER ONE IS FOR EXPOSED PAD VARIATION.

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	21-0106	B 2/2

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